

Problem 1.

- A) It can be demonstrated that the maximum SNR is achieved with an integration window starting in sync with the signal and having a duration $T_G=1.25\tau$. In this case $V_{P,MIN}=7.04\mu V$.
- B) The best solution is sampling the background level integrating only the background for a time interval much longer than the one used for the signal. In this way, the SNR is not affected by the background subtraction.
- C) Since the signal varies slowly over time, a Boxcar is the best solution. A conservative choice is $T_F=12.5ms$ to have a weighting function that goes to almost zero in 1s (considering 5τ). The improvement in the SNR with respect to the single GI is $\sqrt{2 * T_F/T_G} = 44.7$.
- D) See theory.

Problem 2.

- A) Additional filtering is necessary to limit 1/f noise. To this aim we can use a high pass filter with $f_i=1MHz$. to further improve the measurement, also a low pass filter can be used with $f_s=10Hz$. Noise contribution= $6\mu V$, dominated by 1/f component.
Signal: $25mV/4 * \alpha * \Delta T$. Typical value of $\alpha=4*10^{-3}$.
 $\Delta T_{MIN}=6\mu V/(25mV/4 * \alpha) = 0.24^\circ C$.
- B) The best solution is sampling the signal where it has the maximum amplitude (in absolute value). To limit 1/f noise we sample both when the signal is maximum (50 samples) and when the signal is minimum (50 samples) and then we subtract the two measurements. See slides of lesson SR17 (BPF3) from page 16 for more details.
 $S= 25mV/4 * \alpha * \Delta T = 25mV/4 * \alpha * \Delta T$
 $N^2 \approx S_v * 1/(2T_G) * f_{PA}/(2f_s)$, where $T_G=100/(2f_s)$ and f_s is the signal frequency (100KHz).
 $\Delta T_{MIN}= 0.4^\circ C$.
- C) The constraint on signal frequency and bandwidth prevent the use of a resonant filter. Instead, a LIA can be used, with a LPF having a $f_p=10Hz$.
 $S= 25mV/4 * \alpha * \Delta T$
 $N^2= 2*S_v * \pi/2 * f_p$.
 $\Delta T_{MIN}= 0.0045^\circ C$
- D) A PN junction temperature sensor can be used in this case. See theory for more details.